

1 DC OFFSET  
166 CORRECTING

Eqn/Iout  
Eqn Out

SLICER

ISLICER

26 QSLICER

22

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QSYMB

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ISYMB

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QPC OFFSET

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QPC CORRECTION

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NCO

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CARRIER LOOP FILTER

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AVG

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AVG

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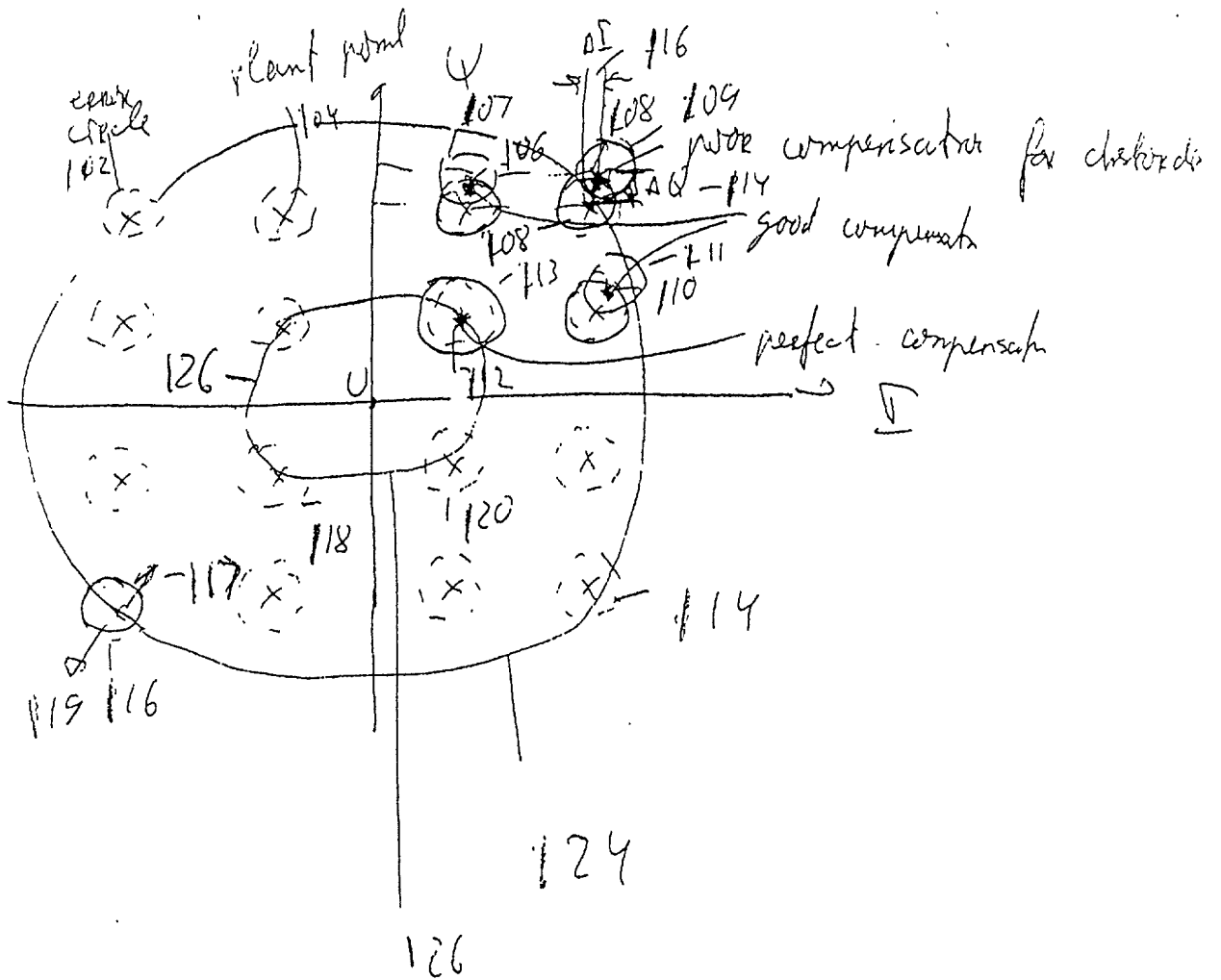
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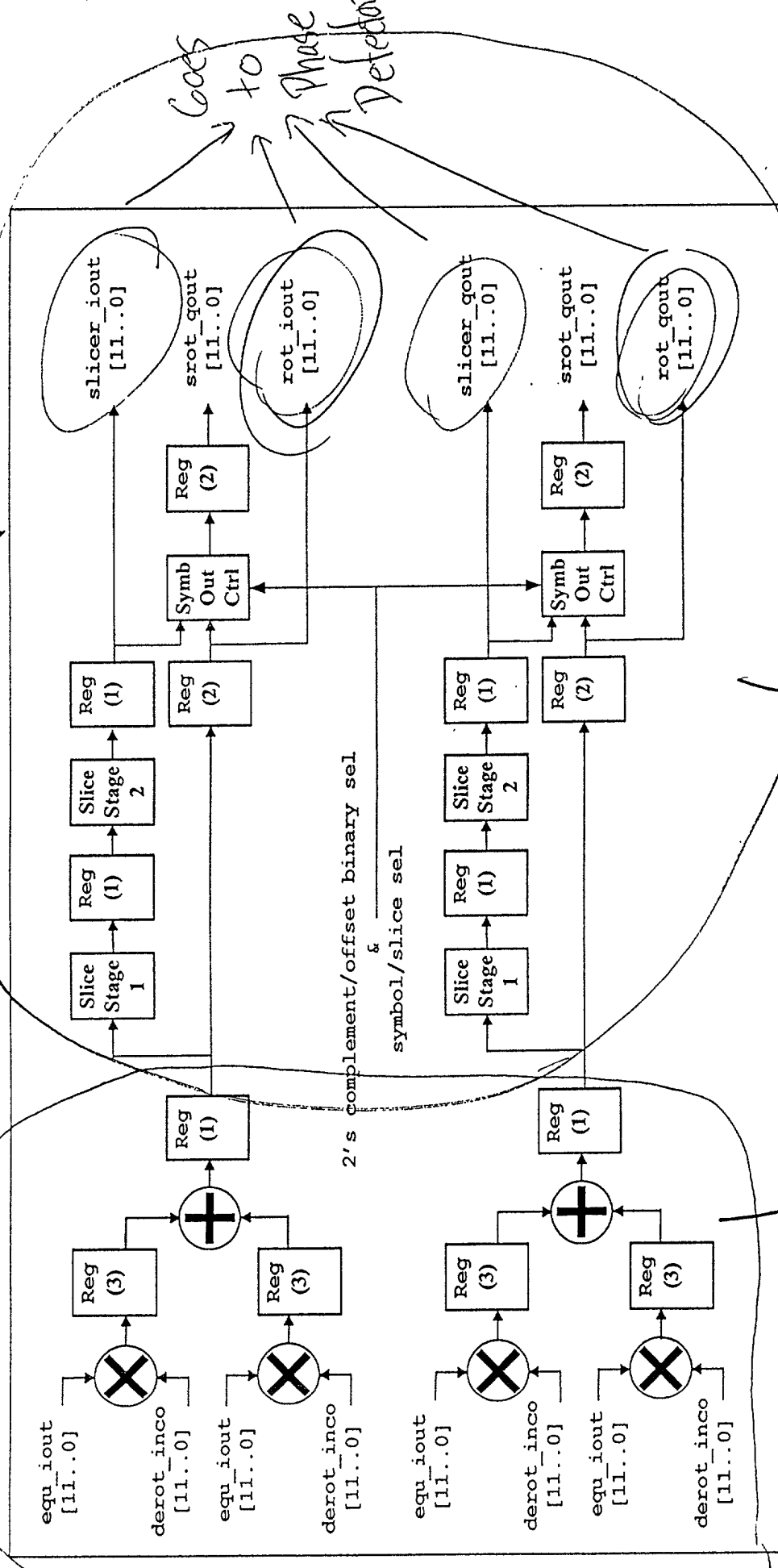
09981085.101601



Module - 105 / Gadh - 195

This is HDL code  
 complex multiplier used in the  
 slicer PLL

Derotator (H)  
 Complex Multiplier

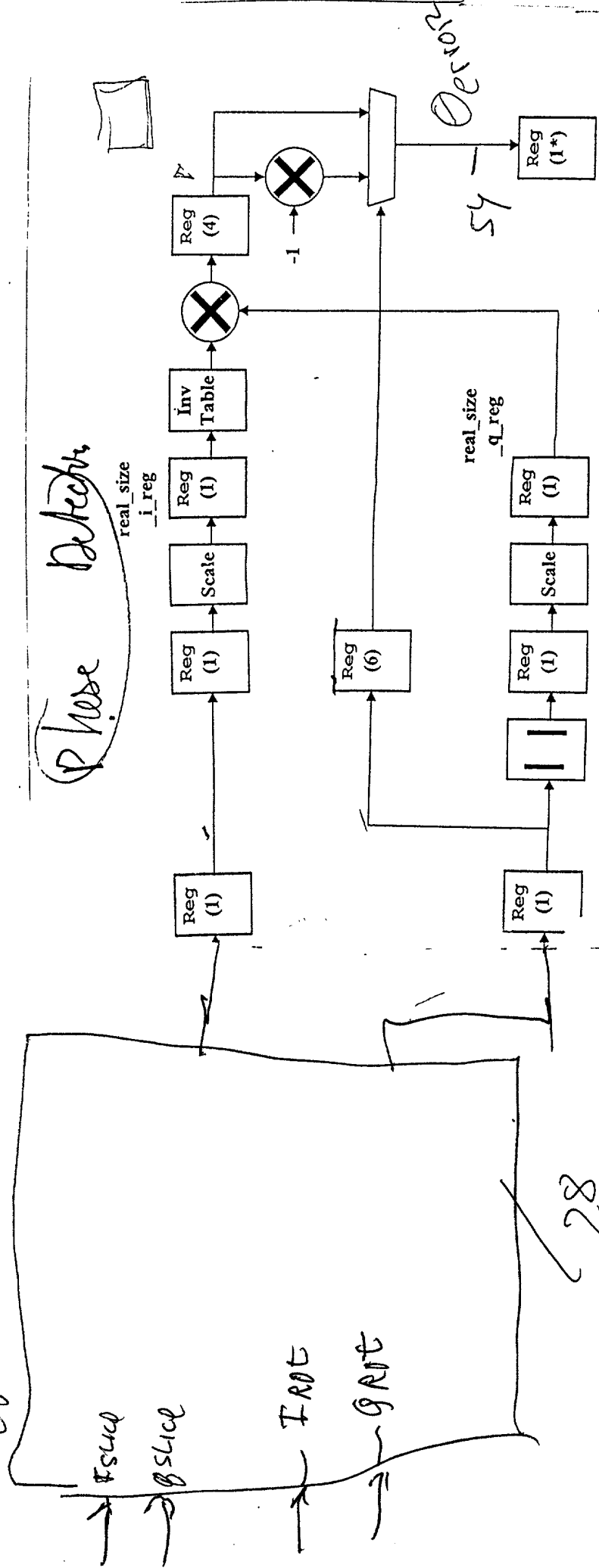


Carrier Derotation 22

16

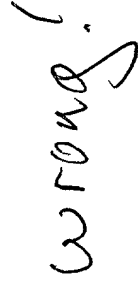
Fig. 3

complex multiplier MISSING (E)  
consgate

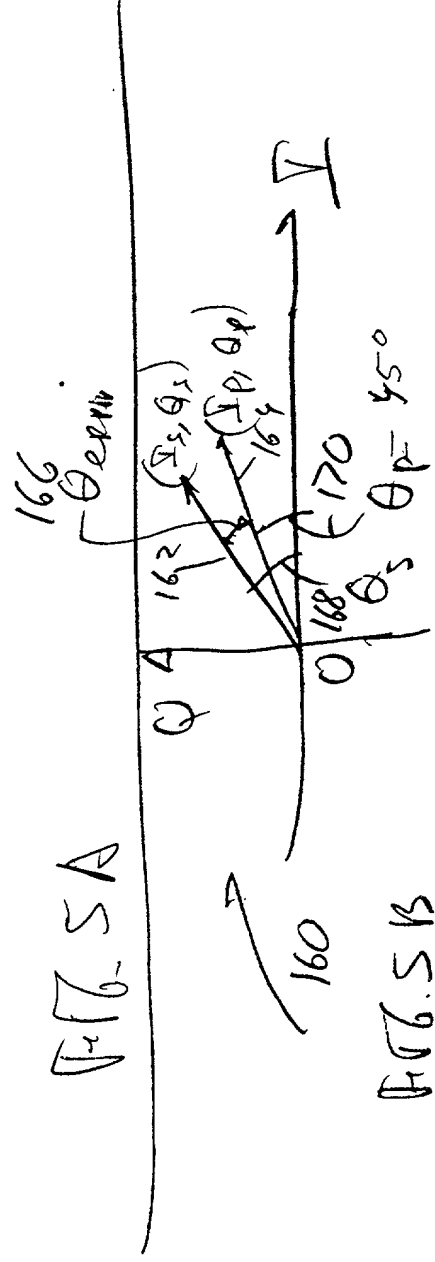


Clocked when  
clk\_valid = 0 and  
clk\_cnt = 1

FIG. 4

[illegible]

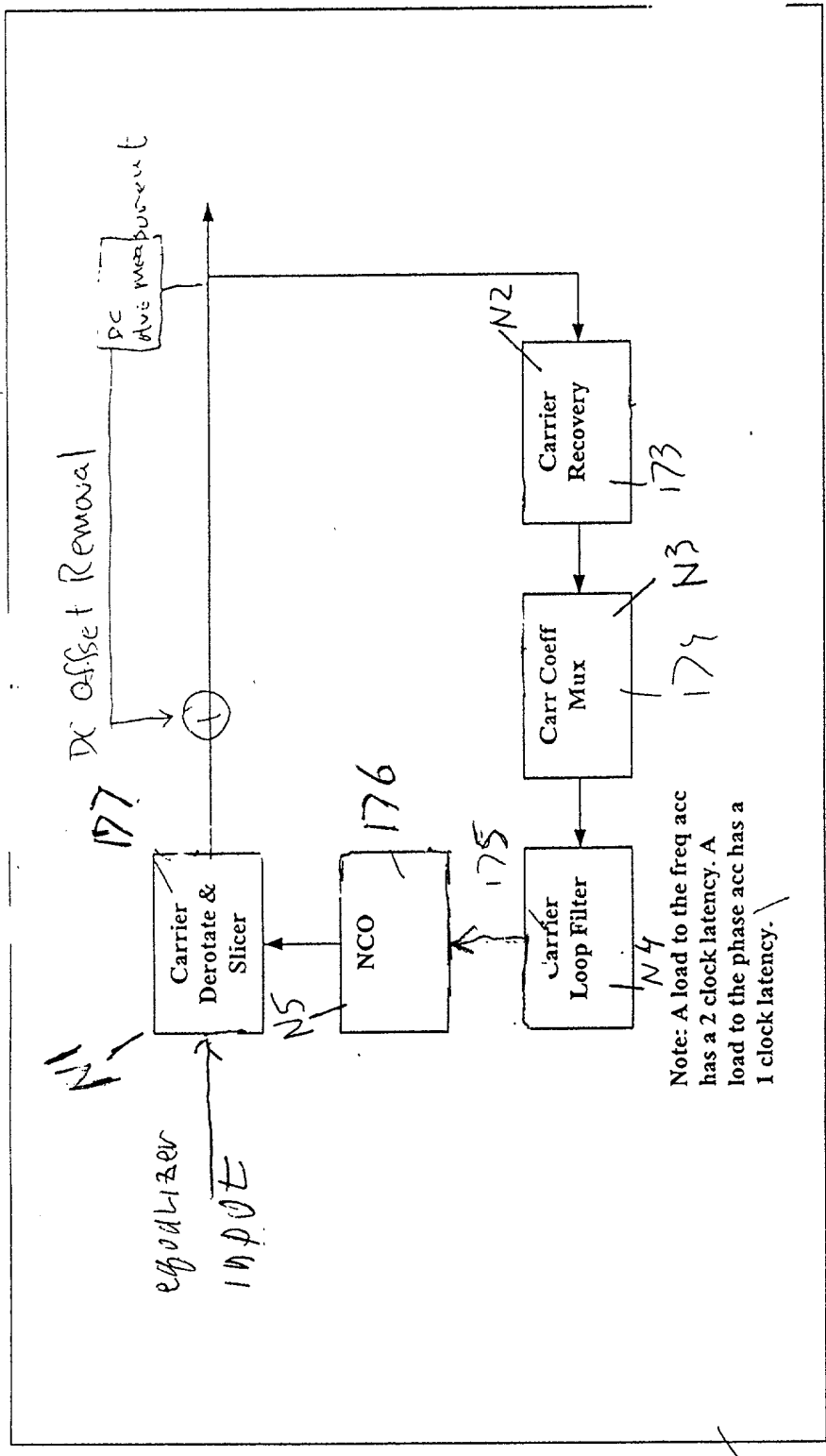
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4599

Worksheet - 105 / Oct-195

TOTOT 58048660



Total Loop Latency

F05C

172

00981085-101601

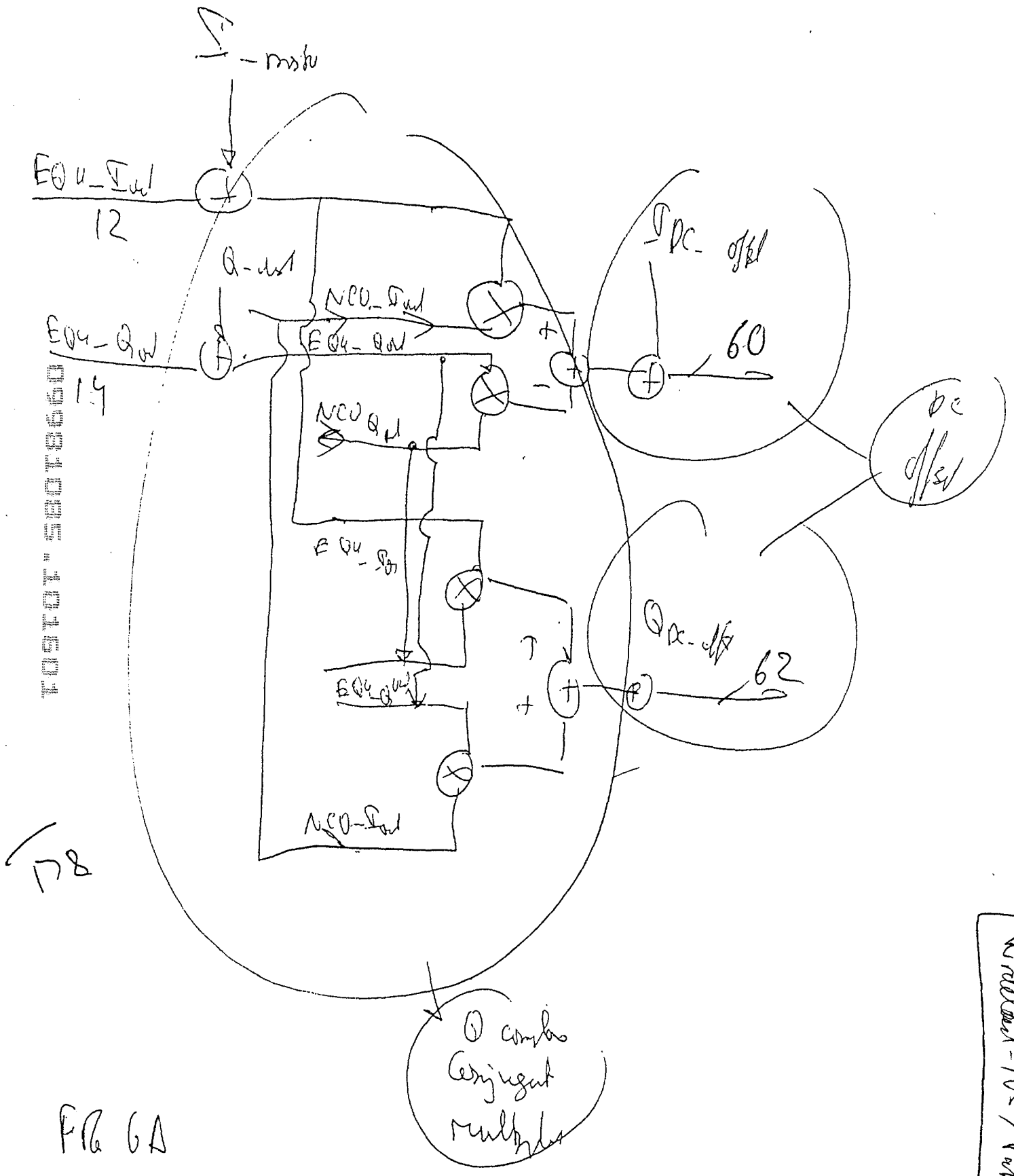


FIG 6A

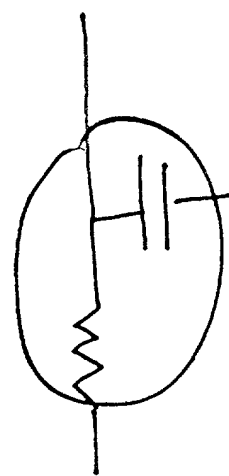
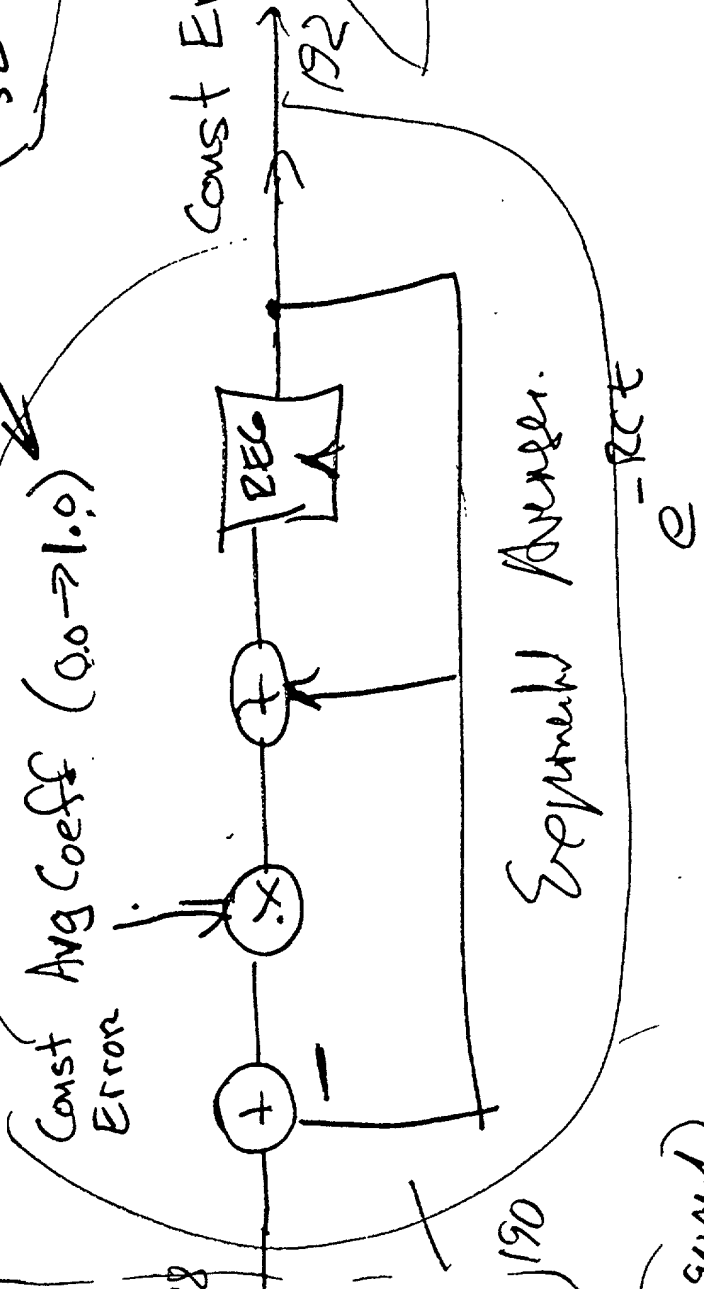
Wideland-105 / Puh-195

Q To direct effect of market power

## Constellation Error Measurement

$\frac{1}{2} N$

~~signals~~  
 Error → 92  
 plants  
~~farm~~  
 rendering  
 noise effect



```

if (Avg Coeff = 0)
    output is held (input ignored)
if (Avg Coeff = 1)
    output = input
else
    output is exponential avg
    of inputs

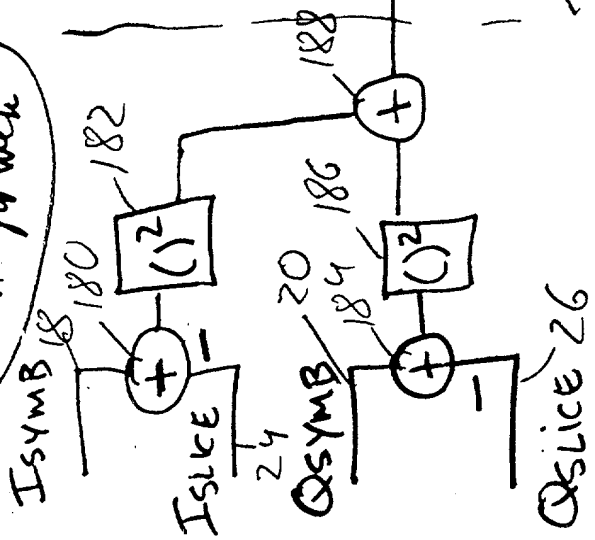
```

PL. 6B

23

Number of  
Error Power Squared.

Insider  
even power  
YMB 18







09031033 104604

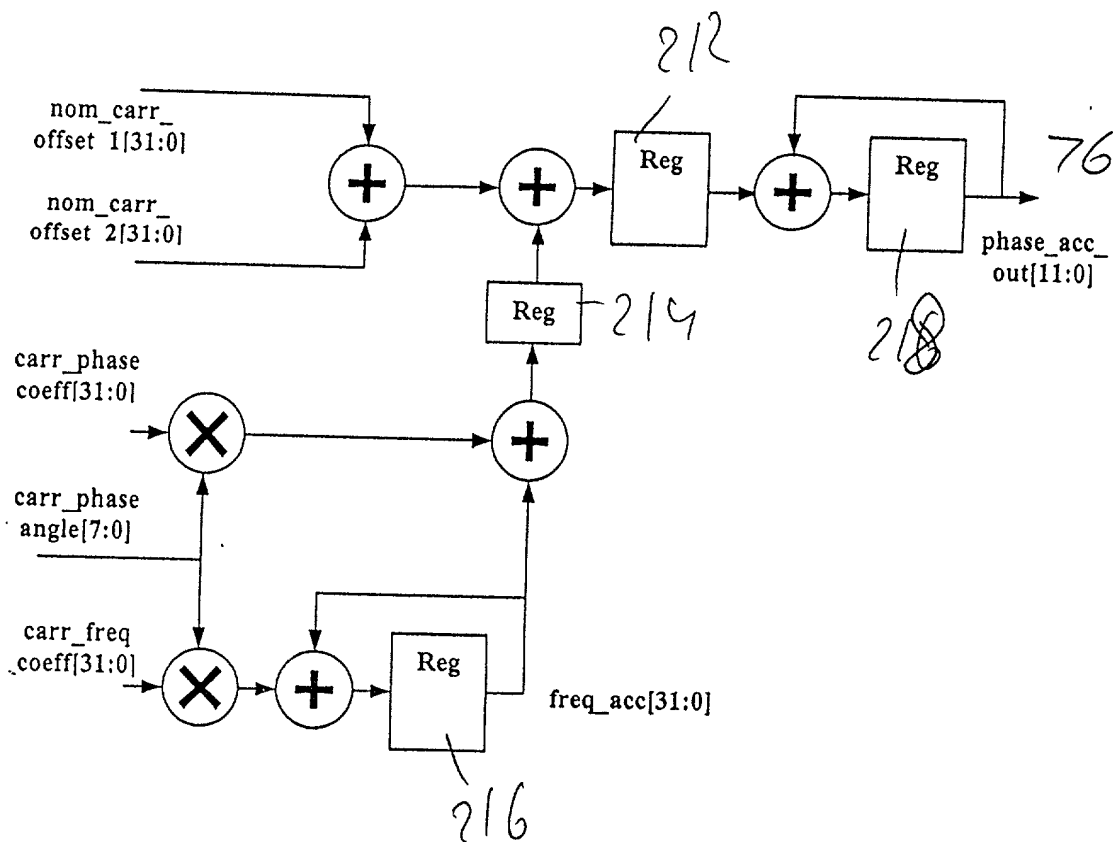


Fig 8. Carrier Loop Filter

231 ~~start~~ 232 250, 258

(A) Sampling a QAM signal received from a transmission channel.

(B) Recovering a symbol clock function from the sampled QAM signal.

(C) Applying the sampled QAM signal to the adaptive equalizer in order to obtain a QAM equalized signal in a Blind Equalization (BE) mode.

(D) Using a slicer to locate a nearest plant point for the QAM BE equalized signal for each recovered symbol clock.

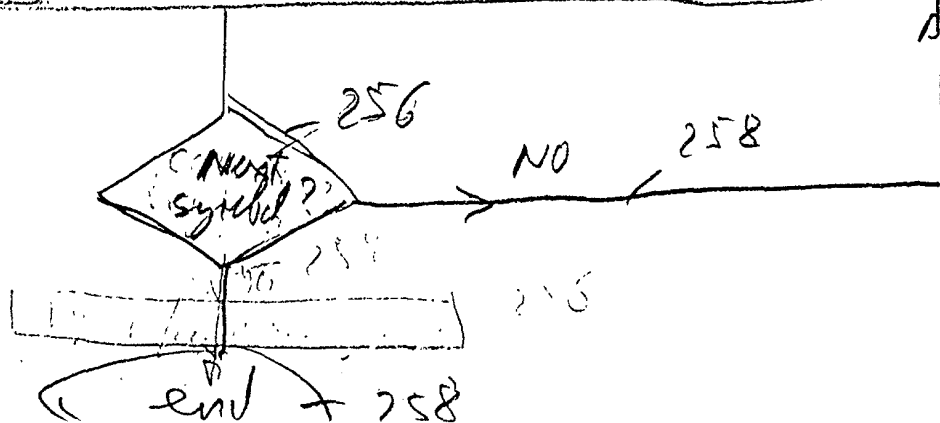
(E) ~~Using a phase detector to obtain~~ an instantaneous inphase component and an instantaneous quadrature component of a phase error signal by comparing an inphase component and a quadrature component of the QAM BE equalized signal and an inphase and a quadrature component of the nearest plant point for each symbol clock.

*Linear phase detector*  
(F) Using a ~~complex conjugate multiplier~~ to translate the inphase component and the quadrature component of the phase error signal into an instantaneous phase ~~error vector~~ *angle* for each symbol clock.

(G) Averaging the instantaneous phase error vector signal by using a carrier loop filter.

(H) Using a complex multiplier to insert an inverse of the averaged phase error vector signal into the QAM BE equalized signal to compensate for the carrier phase error.

(I) Repeating the steps (D-H) to close a carrier frequency loop.



230  
FIG. 9

Updated - 105/6 and - 195

Selecting an initial set of PID coefficients by using the state machine to set the variable bandwidth of the carrier loop filter to be higher than a frequency uncertainty during a QAM signal acquisition state of the QAM demodulator.

262

Adjusting the initially selected set of PID coefficients by using the state machine in order to decrease the initially set bandwidth of the carrier loop filter in incremental stages to be less than the frequency uncertainty during a carrier tracking state of the QAM demodulator.

264

244

Step 6. - Normal Mode  
F16.10

Modela-105/8 at-135

266

(A) Starting with a first set of coefficients of the carrier frequency loop in the state machine corresponding to a normal set of input code words.

(B) Detecting a burst set of input code words.

270 → 272 → 274

(C) Selecting a second set of coefficients of the carrier frequency loop in the state machine corresponding to the burst set of input code words for a predetermined amount of time to switch the QAM modem to a burst mode of operation.

(D) Switching the state machine back so that to set the carrier frequency loop includes the first set of coefficients after the burst mode is over.

280 → 282 → 284

(E) Repeating the steps (A-D).

244

Step 6 - Burst Mode

FIG. 1.1